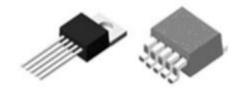


Description

The B2576 series of regulators are monolithic integrated circuit that provides all the active

functions for a step -down (buck) switching regulator , capable of driving 3 A load with excellent line and

load regulation . The B2576 available in fixed output voltages of 3 . 3 V , 5 V , 1 2 Vand an adjustable output version .



TO220T-5L TO263-5L

Requiring a minimum number of external components, these regulators are simple to use and include internal frequency compensation and a fixed-frequency oscillator.

The B2576 series offers a high- efficiency replacement for popular three -terminal linear regulators .

It substantially reduces the size of the heat sink , and in some cases no heat sink is required. A standard series of inductors optimized for use with the B2576 are available from several different manufacturers . This feature greatly simplifies the design of switch- mode power supplies .

Other features include a guaranteed ± 4 % tolerance on output voltage within specified input voltages and output load conditions, and ± 1.0 % on the oscillator frequency.

External shutdown is included , featuring 5 0 μ A (typical) standby current . The output switch includes cycle -by - cycle current limiting , as well as thermal shutdown for full protection under fault conditions .

The B2576 is available in TO220T - 5 L and TO263 - 5 L package .

Features

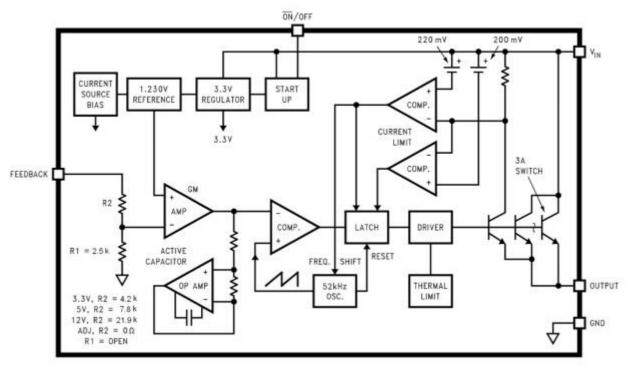
- \succ 3.3 V, 5 V, 12 V and adjustable output versions
- ➢ High efficiency
- Guaranteed 3 A output current
- Requires only 4 external components
- ➢ 5 2 kHz fixed frequency internal oscillator
- > TTL shutdown capability , low power standby mode
- > Uses readily available standard inductors
- > Thermal shutdown and current limit protection
- Adjustable version output voltage range, 1.23 V to 37 V ± 4% max over line and load conditions



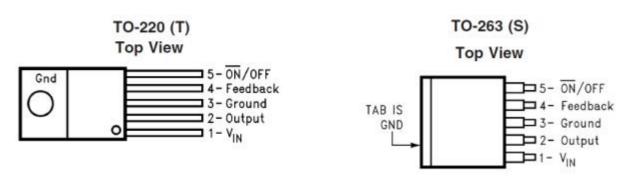
Application

- Simple high- efficiency step -down (buck) regulator
- Efficient pre regulator for linear regulators
- On card switching regulators
- > Positive to negative converter (Buck-Boost)

Block Diagram



Pin Configuration (Top View)





Pin Description

Pin Number	Pin Name	Function Description
1	VIN	This is the positive input supply for the IC switching regulator. Bypass Vin to GND with a suitably large capacitor to eliminate noise on the input.
2	OUTPUT	Internal switch, the voltage at this pin switches between (+V $_{\rm IN}$ V $_{SAT})$ and approximately -0.5V
3	GROUND	Circuit ground Care must be taken in layout. This pin should be placed outside of the Schottky Diode to output capacitor ground path to prevent switching current spikes from inducing voltage noise into B2576
4	FEEDBACK	Senses the regulated output voltage to complete the feedback loop The feedback threshold voltage is 1.23V.
5	ON/ OFF	Allows the switching regulator circuit to be shut down using logic level signals

Absolute Maximum Ratings

Parameter Name	Symbol	Value	Unit
Maximum Supply Voltage	V IN	45	V
Pin Input Voltage	ON/ OFF	-0.3V≤V≤+V _{IN}	V
Output Voltage to Ground(steady state)	V _{OUT}	- 1	V
Power Dissipation	P _{DMAX}	Internally Limited	
Thermal Resistance (TO220 & TO263)	Du	50	°C/W
(Junction to Ambient, No Heatsink, Free Air)	Rja	50	U/ VV
Storage Temperature Range	Tstg	-65~+150	°C



Maximum Junction Temperature	T _{JA}	150	°C
Lead Temperature (Soldering, 10 sec)	T _{LEAD}	260	°C
ESD Susceptibility (Human Body Model)	ESD	2000	V

Recommended Operating Conditions

Parameter Name	Symbol	Value	Unit
Supply Voltage	VIN	8~40	V
Operating Temperature Range	Topr	-40~+125	°C

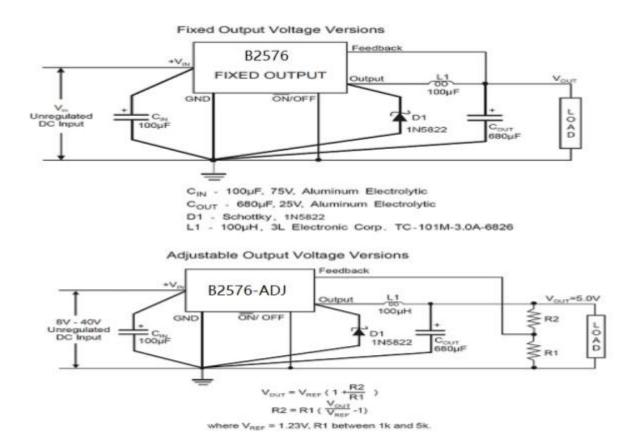
Electrical Characteristics(Unless otherwise specified: $T_J = 25 \degree C$)

Parameter Name	Symbol	Test Conditions	Min	Тур	Max	Un its
Device Parameters			· ·			
Feedback Bias Current	lb	Adjustable version only, $V_{OUT} = 5V$		50	100	nA
Oscillator Frequency	fo	(Note 1)	47	52	58	kH z
VSAT Saturation Voltage	VSAT	I _{OUT} =3A		1.4	1.8	V
Max. Duty Cycle (ON)	DC		93	98		%
Current Limit	I _{CL}		3.2	3.5	4.5	Α
Output Leakage Current		Output = 0 V			2	mA
Sulput Leakage Sullent	Ι _L	Output = - 1 V		7.5	30	mA
Quiescent Current	l _Q			5	10	mA
Standby Quiescent Current	S T BY	pin=5V(OFF)		50	200	μA
ON/ OFF Control						
	VIH	V OUT = 0 V	2.0			V
<i>ON/ OFF</i> Pin Logic Input Level	V _{IL}	V ou⊤=nominal output voltage			0.8	v
	IIH	\overline{ON}/OFF pin = 5 V (OFF)		12	30	μA
ON/ OFFPin Input Current	I _{IL}	\overline{ON}/OFF pin = 0 V (ON)		0	10	μA
B2576-3.3V						
		$V_{IN} = 12V, I_O = 500 \text{mA}$	3.234	3.3	3.366	V
Output Voltage	V _{OUT}	8V≤V _{IN} ≤40V 0.5A≤1load≤3A	3.168	3.3	3.432	v
Efficiency	η	VIN=12V, ILOAD=3A		75		%
B2576-5.0V						



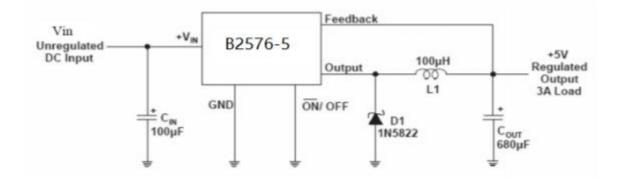
		$V_{IN} = 12V, I_O = 500 \text{ mA}$	4.900	5.000	5.100	V
Output Voltage	V _{OUT}	8V≤VIN≤40V 0.5A≤Iload≤3A	4.800	5.000	5.200	V
Efficiency	η	V _{IN} =12V, I _{LOAD} =3A		77		%
B2576-12V			L			
		$V_{IN} = 18V, I_O = 500 \text{ mA}$	11.76	12.000	12.24	V
Output Voltage	V _{OUT}	15V≤V _{IN} ≤40V 0.5A≤Iload≤3A	11.52	12.000	12.48	V
Efficiency	η	V _{IN} =18V, I _{LOAD} =3A		88		%
B2576- ADJ		1				
		V _{IN} =12V,I _O =500mA, V _{OUT} =5V	1.217	1.230	1.243	V
Output Voltage	V _{OUT}	$8V \le V_{IN} \le 40V,$ $V_{OUT} = 5V$ $0.5A \le 1$ LOAD $\le 3A$	1.193	1.230	1.267	V
Efficiency	η	V _{IN} =12V,I _{LOAD} =3A, V _{OUT} =5V		77		%

Test Circuit





Typical Application

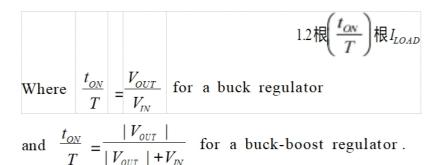


Application Information Input Capacitor (C_{IN})

To maintain stability, the regulator input pin must be bypassed with at least a 100 μ F electrolytic capacitor. The capacitor's leads must be kept short, and located near the regulator. If the operating temperature range includes temperatures below -25°C, the input capacitor value may need to be larger. With most electrolytic capacitors, the

capacitance value decreases and the ESR increases with lower temperatures and age.

Paralleling a ceramic or solid tantalum capacitor will increase the regulator stability at cold temperatures. For maximum capacitor operating lifetime, the capacitor's RMS ripple current rating should be greater than



Inductor s election

All switching regulators have two basic modes of operation: continuous and

discontinuous. The difference between the two types relates to the inductor current, whether it is flowing continuously, or if it drops to zero for

a period of time in the normal switching cycle. Each mode has distinctively different

operating characteristics, which can affect the regulator performance and requirements. The B2 576 can be used for both continuous and discontinuous modes of operation.

When using inductor values shown in the inductor selection guide, the peak-to-peak inductor ripple current will be approximately 20% to 30% of the maximum DC current.



With

relatively heavy load currents, the circuit operates in the continuous mode (inductor current always flowing), but under light load conditions, the circuit will be forced to the discontinuous mode (inductor current falls to zero for

a period of time). This discontinuous mode of operation is perfectly acceptable. For light loads (less than approximately 300mA) it may be desirable to operate the

regulator in the discontinuous mode, primarily because of the lower inductor values required

for the discontinuous mode. The selection guide chooses inductor values suitable for

continuous mode operation, but if the inductor value chosen is

prohibitively high, the designer should investigate the possibility of discontinuous operation.

Inductors are available in different styles such as pot core, toriod, E-frame, bobbin

core, etc., as well as different core materials, such as ferrites and powdered iron. The least expensive, the bobbin core type, consists of wire wrapped on a ferrite rod core.

This type of construction makes for an inexpensive inductor, but since the magnetic flux is not completely contained within the core, it generates more electromagnetic interference (EMI). This EMI can cause problems in sensitive circuits, or can give incorrect scope readings because of induced voltages in the scope probe. The inductors listed in the selection chart include ferrite pot core

construction for A $\ensuremath{\mathsf{IE}}$, powdered iron toroid for Pulse Engineering , and ferrite bobbin core for Ren co .

An inductor should not be operated beyond its maximum rated current because it may saturate . When an inductor begins to saturate , the inductance decreases rapidly and the inductor begins to look mainly resistive (the DC resistance of the winding) This will cause the switch current to rise very rapidly . Different inductor types have different

saturation characteristics, and this should be kept in mind when selecting an inductor. The inductor manufacturer 's data sheets include current and energy limits to avoid inductor saturation.

Inductor Ripple Current

When the switcher is operating in the continuous mode, the inductor current waveform ranges from a triangular to a sawtooth type of waveform (depending on the input

voltage). For a given input voltage and output voltage, the peak-to-peak amplitude of this inductor current waveform remains constant. As the load current rises or falls, the entire sawtooth current waveform also rises or falls. The average DC value of this

waveform is equal to the DC load current (in the buck regulator configuration). If the load current drops to a low enough level, the bottom of the sawtooth current waveform will reach zero, and the switcher will change to a discontinuous mode of operation.

This is a perfectly acceptable mode of operation. Any buck switching regulator (no

matter how large the inductor value is) will be forced to run discontinuous if the load current is light enough.



Catch Diode

Buck regulators require a diode to provide a return path for the inductor current when the switch is off. This diode

should be located close to the B2576 using short leads and short printed circuit traces. Because of their fast switching speed and low forward voltage drop, Schottky diodes provide the best

efficiency, especially in low output voltage switching regulators (less than 5V).

Fast-Recovery, High-Efficiency, or Ultra-Fast Recovery diodes are also suitable, but some types with an abrupt turn-off characteristic may cause instability and EMI

problems. A fast-recovery diode with soft recovery characteristics is a better choice. Standard 60 Hz diodes (e.g., 1N4001 or 1N5400, etc.) are also not suitable.

Output Capacitor

An output capacitor is required to filter the output voltage and is needed for loop stability. The capacitor should be located near the D2576E using short pc board traces. Standard aluminum electrolytics are usually adequate, but low ESR types are

recommended for low output ripple voltage and good stability. The ESR of a capacitor depends on many factors, some which are: the value, the voltage rating, physical size and the type of construction. In general, low value or low voltage (less than 12V) electrolytic capacitors usually have higher ESR numbers.

The amount of output ripple voltage is primarily a function of the ESR (Equivalent Series Resistance) of the

output capacitor and the amplitude of the inductor ripple current (ΔI_{IND}) . See the section on inductor ripple current in Application Hints. The lower capacitor values $(220 \ \mu F - 1000 \ \mu F)$ will allow typically $50 \ mV$ to $150 \ mV$ of output ripple voltage, while largervalue capacitors will reduce the ripple to approximately $20 \ mV$ to $50 \ mV$. Output Ripple Voltage = (ΔI_{IND}) (ESR of C_{OUT}). To further reduce the output ripple voltage, several standard electrolytic capacitors may be paralleled, or a higher-grade capacitor may be used. Such capacitors are often called "high-frequency" "low-inductance" or "low-ESR" These will reduce the output ripple to $10 \ mV$ or $20 \ mV$. However, when

operating in the continuous mode, reducing the ESR below 0.03Ω can cause instability in the regulator. Tantalum capacitors can have a very low ESR, and should be

carefully evaluated if it is the only output capacitor. Because of their good low temperature characteristics, a tantalum can be used in parallel with aluminum

electrolytics, with the tantalum making up 10% or 20% of the total capacitance. The capacitor 's ripple current rating at 52kHz should be at least 50% higher than the

peak-to-peak inductor ripple current.



Output Voltage Ripple and Transients

The output voltage of a switching power supply will contain a sawtooth ripple voltage at the switcher frequency, typically about 1% of the output voltage, and may also

contain short voltage spikes at the peaks of the sawtooth waveform. The output ripple voltage is due mainly to the inductor sawtooth ripple current multiplied by the ESR of the output capacitor. The voltage spikes are present because of the fast switching

action of the output switch, and the parasitic inductance of the output filter capacitor. To minimize these voltage spikes, special low inductance capacitors can be used, and their lead lengths must be kept short. Wiring inductance, stray capacitance, as well as the scope probe used to evaluate these transients, all contribute to the amplitude of

these spikes. An additional small LC filter $(20 \mu H \& 100 \mu F)$ can be added to the output to further reduce the amount of output ripple and transients. A 10 x reduction in

output ripple voltage and transients is possible with this filter.

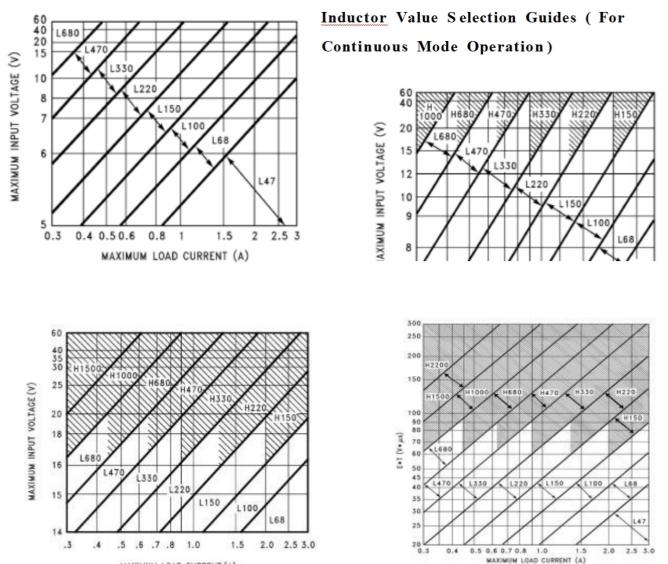
Feedback Connection

The B2576 (fixed voltage versions) feedback pin must be wired to the output voltage point of the switching power supply. When using the adjustable version, physically locate both output voltage programming resistors near the B2576 to avoid picking up unwanted noise. Avoid using resistors greater than $100k\Omega$ because of the increased chance of noise pickup.

ON/ OFF Input

For normal operation, the ON/OFF pin should be grounded or driven with a low-level TTL voltage (typically below 1.6V). To put the regulator into standby mode, drive this pin with a high-level TTL or CMOS signal. The to $+V_{IN}$ without a resistor in series with it. The be left open.





Inductor Code	Inductor Value	Schott	Pulse Eng.	Ren co
L47	47 µ H	671 26980	PE-53112	RL2442
L68	68 µ H	671 26990	PE-92114	RL2443
L100	100 µ H	671 27000	PE-92108	RL2444
L150	150 µ H	671 270 10	PE-53113	RL1954
L220	220 µ H	671 27020	PE-52626	RL1953
L330	330 µ Н	671 27030	PE-52627	RL1952
L470	470 µ H	671 27040	PE-53114	RL1951
L680	680 µ H	671 27050	PE-52629	RL1950
H150	150 µ H	671 27060	PE-53115	RL2445
H220	220 µ Н	671 27070	PE-53116	RL2446
Н330	330 μ Н	671 27080	PE-53117	RL2447
H470	470 µ H	671 27090	PE-53118	RL1961
H680	680 µ H	671 27100	PE-53119	RL1960

MAXIMUM LOAD CURRENT (A)



H1000	1000 µ Н	671 27110	PE-53120	RL1959

H1500	1500 µ H	671 27120	PE-53121	RL1958
H2200	2200 μ Н	671 27130	PE-53122	RL2448

VR	Sc	Schottky		ecovery
	ЗA	4A-6A	3A	4A-6A
20V	1N5820	1N5823		
	MBR320P			
	SR302			
30V	1N5821	50WQ03	1	
	MBR330	1N5824		
	31DQ03		1.000 200 x 114 x	The following
	SR303	The following	-	diodes are all
40V	1N5822	MBR340	diodes are all rated to 10	rated to 100V
	MBR340	50WQ04	rated to 100V	
	31DQ04	1N5825	31DF1	50WF10
	SR304		HER302	MUR410
50V	MBR350	50WQ05		HER602
	31DQ05			
	SR305			
60V	MBR360	50WR06	1	
	DQ06	50SQ060		
	SR306			

Inductor Selection Guide

Diode Selection Guide



Outline Dimensions

TO-220T-5L

Unit: mm



